7.5 Amp LDO Adjustable Positive Voltage Regulators

# VRG8691/92

### Features

- Radiation performance
  - Total dose: 100 krad(Si),

Dose rate = 50-300 rad(Si)/s

- Output voltage adjustable: 1.0V to 3.3V
- Output current: 7.5A
- Dropout voltage: 0.5V at 7.5Amps
- Voltage reference: 1.0V ±0.5%
- Load regulation: 0.5% max
- Line regulation: 0.2% max
- Ripple rejection: >80dB
- Enable Input TTL / CMOS Compatible
- Slow Start capability
- · Stable with multiple ceramic output capacitors
- Packaging Hermetic metal
  - Thru-hole or Surface mount
  - 12 Leads, 0.900"L x 1.000"W x .205"Ht
  - Power package
  - Weight 18 gm max
- Designed for aerospace and high reliability space applications
- Radiation Hardness Assurance Plan: DLA Certified to MIL-PRF-38534, Appendix G.

### Description

The VRG8691/92 is capable of supplying in excess of 7.5Amps over the output voltage range as defined under recommended operating conditions. The regulator is exceptionally easy to set-up, requiring only 2 external resistors to set the output voltage. The module design has been optimized for excellent regulation and low drop-out voltage. Figures 2 through 5 illustrate setting output voltage, setting current limits and choosing a slow start capacitor. The VRG8691/92 serves a wide variety of applications including local on-card regulation, programmable output voltage regulation or precision current regulation.

The VRG8691/92 has been specifically designed to meet exposure to radiation environments. The VRG8691 is configured for a Thru-Hole 12 lead metal power package and the VRG8692 is configured for a Surface Mount 12 lead metal power package. It is guaranteed operational from -55°C to +125°C. Available screened to MIL-STD-883, the VRG8691/92 is ideal for demanding military and space applications.

### Current Limit (I<sub>CL</sub>)

The VRG8691/92 features internal current limiting making them virtually blowout-proof against overloads. The limit is nominally  $11.5A \otimes Vin = 5V$  (see Table 2), but may be increased or decreased with the addition of one external resistor (see Application Note 2).



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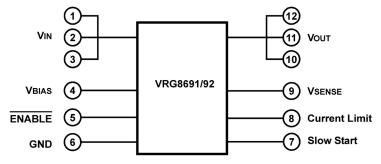


Figure 1 – Block Diagram / Schematic

### **Absolute Maximum Ratings**

Parameter	Range	Units
Operating (Junction) Temperature Range	-55 to +150	°C
Lead Temperature (soldering, 10 sec)	300	°C
Storage Temperature Range	-65 to +150	°C
V <sub>BIAS</sub> , V <sub>IN</sub>	7	V
Thermal Resistance (Junction to case $\Theta_{JC}$ )	1	°C/W
Power	25 <u>1</u> /	W

<u>1</u>/Based on pass transistor limitations of ( $V_{IN}$  -  $V_0$ ) x I<sub>0</sub> and  $\Theta_{JC}$  < 1°C/W with 25°C max T<sub>J</sub> rise and T<sub>C</sub> = +125°C.

#### Notice:

 Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only; functional operation beyond the "Operation Conditions" is not recommended and extended exposure beyond the "Operation Conditions" may effect device reliability.

### **Recommended Operating Conditions**

Parameter	Range	Units
Output Voltage Range	1.0 to 3.3	V <sub>DC</sub>
Case Operating Temperature Range	-55 to +125	°C
Output Current	0 to 7.5	А
V <sub>BIAS</sub>	3.3 to 5.5 <u>1</u> /	V <sub>DC</sub>
V <sub>IN</sub>	1.8 to 5.5 <u>2</u> /	V <sub>DC</sub>

Notes:

- 1)  $V_{\text{BIAS}}$  must maintain a level equal or above  $V_{\text{IN}}$  but not fall below 3.3V
- 2) Depending upon VOUT setting.



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### Electrical Performance Characteristics 1/

Parameter	SYM	Conditions	MIN	MAX	Units
Reference Voltage	V <sub>REF</sub>	$V_{IN} = V_{BIAS} = 5V$ , $\overline{ENABLE} = 0$ , $0A \le I_{OUT} \le 7.5A$	0.985	1.015	V
Line Regulation	$\frac{\Delta V_{OUT}}{\Delta V_{IN}}$	$\begin{array}{l} 2V \leq V_{IN} \leq 3V,  V_{OUT} = 1.0V,  C_{IN} \geq 47 \mu \text{F}, \\ 4.3V \leq V_{IN} \leq 5.3V,  V_{OUT} = 3.3V,  C_{OUT} \geq 47 \mu \text{F}, \end{array}$	-	0.2	%/V
Load Regulation	$\frac{\Delta V_{OUT}}{\Delta I_{OUT}}$	$0A \le I_{OUT} \le 7.5A, C_{IN} \ge 47\mu F, C_{OUT} \ge 47\mu F,$	-	0.5	%
Ripple Rejection Ratio		$      f = 120Hz, C_{LOAD} = 47 \mu F, \\ V_{IN} + V_{RIP} \ge V_{OUT} + V_{DROP(MAX)} @ 5A, \\ V_{IN} = 4.3V, V_{RIP} = 1V_{P-P}, V_{OUT} = 3.3V $	80	-	dB
Dropout Voltage	V <sub>DROP</sub>	$@\Delta V_{OUT} = 1\%, 0A \le I_{OUT} \le 7.5A$	-	0.5	V
Adjustment Pin Current	I <sub>ADJ</sub>		-	100	pА
Minimum Load Current	I <sub>MIN</sub>		-	0	mA
Current Limit <u>2</u> /	I <sub>CL</sub>		9.5	13.5	А
Long Term Stability <u>3</u> /	$\frac{\Delta V_{OUT}}{\Delta T_{IME}}$		-	1	%
Supply Current (V <sub>BIAS</sub> )	I <sub>BIAS</sub>		-	15	mA

#### Notes:

- 1) Unless otherwise specified, these specifications apply for post radiation:  $V_{BIAS} = V_{IN} = 5V$ ,  $V_{OUT} = 3.3V$ ,  $I_{OUT} = 7.5A$  and  $55^{\circ}C < T_{C} < +125^{\circ}C$ , Min Input/Output capacity of  $47\mu$ F Tant with  $1\mu$ F ceramic in parallel.
- 2) Current Limit is adjustable as shown in Application Note 2, Figures 3 and 4.
- 3) Not tested. Shall be guaranteed to the specified limits after 1000hr life test.

### Application Note 1 Basic Set-Up

Setting the output voltage (V<sub>OUT</sub>):



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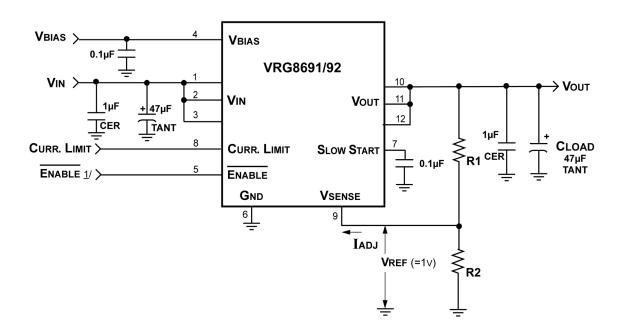


Figure 2 –Setting Output Voltage



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To set the output voltage for a particular  $V_{\mbox{\scriptsize OUT}}$  :

- Choose an R2 value. (Recommended value =  $1k\Omega$ )
- Then use the following formula to determine the value of R1.

$$\label{eq:R1} R1 = R2 \ \times \ \ \frac{V_{OUT} \ - \ V_{REF}}{(R2 \ \times \ I_{ADJ}) + \ V_{REF}} \quad , \mbox{ where } V_{REF} = 1_V, \ I_{ADJ} \ typ = 10 pA$$

Table 1 shows example values for R1 and R2 to achieve some standard voltages.

Table 2 shows the nominal current limit settings if the ' $C_{URR}$ .  $L_{IMIT}$ ' function (pin 8) is left open.

### Table 1

Example R1 & R2 for typical VOUT

Vout	R2	R1
3.3V	1kΩ	2.3kΩ
2.5V	1kΩ	1.5kΩ
1.8V	1kΩ	800Ω
1.0V	1kΩ	0Ω

#### Table 2

#### <u>2/</u>

V <sub>IN</sub>	I <sub>CL NOM</sub>
5V	11.5A
3.3V	7.5A
2.5V	5.7A
1.8V	4.1A

#### Notes:

1)  $\overline{\text{ENABLE}}$  should be asserted after both  $V_{\text{IN}}$  and  $V_{\text{BIAS}}$  are applied.

(See Application Note 3, Figure 5 for the configuration where a separate ENABLE control line is NOT required).

2)  $I_{\text{CL}}$  varies directly with  $V_{\text{IN}}.$ 

(See Application Note 2 for adjusting the Current Limit,  $\ensuremath{I_{\text{CL}}}\xspace$ ).



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### Application Note 2 Setting the Current Limit

To Increase the Current Limit ( $I_{CL}$ ):

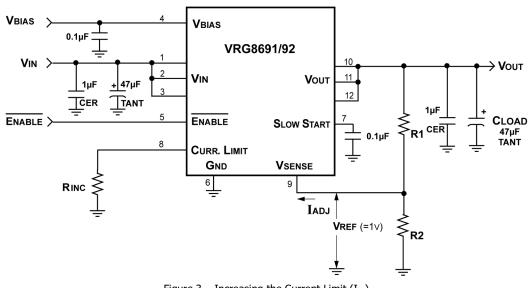


Figure 3 – Increasing the Current Limit (I $_{\mbox{\scriptsize CL}}$ )

- If the 'C\_URR. L\_IMIT' function (pin 8) is left open, the I\_CL decreases from 11.5 A\_(NOM) as V\_IN is decreased from 5V (Table 3).

### Table 3

V <sub>IN</sub>	R <sub>INC</sub>	I <sub>CL NOM</sub>
5V	Open	11.5 A
3.3V	Open	7.5 A
2.5V	Open	5.7 A
1.8V	Open	4.1 A



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- To increase the current limit above the nominal setting for any  $V_{\mbox{\scriptsize IN}}$  and  $I_{\mbox{\scriptsize CL}}$  combination, use the following formula:

$$R_{INC(K\text{-}OHMS)} = \frac{30 \times V_{IN}}{\left(\frac{30 \times I_{CL}}{69}\right) - V_{IN}}$$

- To maintain  $I_{CL}$  at the 11.5 A setting, for commonly found  $V_{IN}$  voltages, apply  $R_{INC}$  value found in Table 4.

### Table 4

V <sub>IN</sub>	R <sub>INC</sub>	I <sub>CL NOM</sub>
5V	Open	11.5 A
3.3V	56kΩ	11.5 A
2.5V	30kΩ	11.5 A
1.8V	16kΩ	11.5 A

### Application Note 2 (continued) Setting the Current Limit

To Decrease the Current Limit (I<sub>CL</sub>):

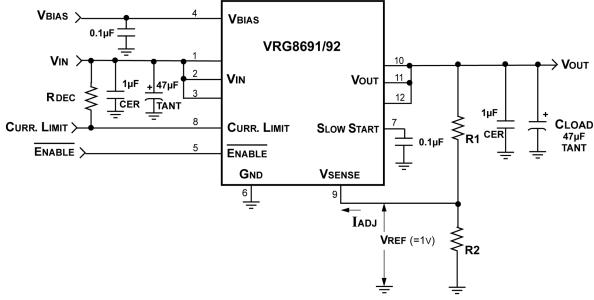


Figure 4 – Decreasing the Current Limit ( $I_{CL}$ )

- As shown in Table 3, if the 'Curr. LIMIT' function (pin 8) is left open, the I<sub>CL</sub> decreases from 11.5 A(NOM) as VIN is decreased from 5V.
- To achieve any  $I_{\text{CL}}$  less than nominal, use  $R_{\text{DEC}}$  which can be calculated using the following formula:

$$R_{\text{DEC(K-OHMS)}} = \frac{31 \times V_{\text{IN}}}{\left(\frac{69 \times V_{\text{IN}}}{30}\right) \text{- I}_{\text{CL}}}$$



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### Application Note 3 Start Up Sequence

### **Recommended Power Supply Sequencing Options:**

- OPTION 1: Controlling the ENABLE line with a Digital signal (TTL / CMOS compatible).
  - Prior to applying power, disable the regulator by setting the ENABLE control line to a HIGH state.
  - Apply VIN and VBIAS. 1/
  - Wait until both  $V_{\ensuremath{\text{IN}}}$  and  $V_{\ensuremath{\text{BIAS}}}$  supplies have reached their operating levels.
  - Toggle the  $\overline{\text{ENABLE}}$  control line to a LOW state to turn on  $V_{\text{OUT}}$  of the regulator.
- OPTION 2: Controlling the  $\overline{\text{ENABLE}}$  line using the  $C_{\text{DELAY}}$  feature.
  - Connect a  $C_{\text{DELAY}}$  capacitor between  $V_{\text{BIAS}}$  and the  $\overline{\text{ENABLE}}$  as shown in Figure 5 below. 2/
  - Apply  $V_{\text{IN}}$  and  $V_{\text{BIAS}}.$   $\underline{1}/$
  - $C_{\text{DELAY}}$  causes the regulator to self-enable after  $V_{\text{BIAS}}$  has reached operating level.

#### Note:

1) The  $\overline{\text{ENABLE}}$  should always be asserted AFTER  $V_{IN}$  and  $V_{BIAS}$  have reached operating level.

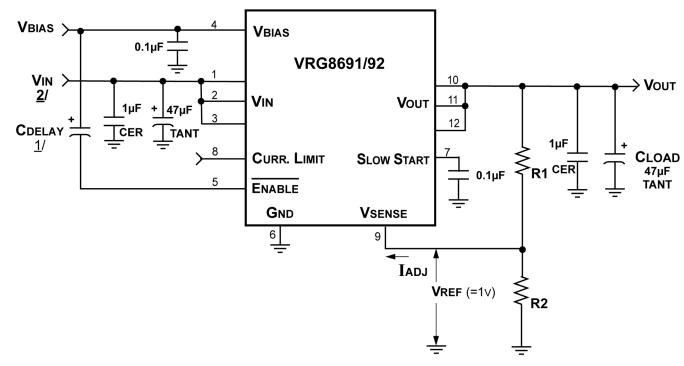


Figure 5 – Delayed Enable

#### Notes:

- 1)  $V_{\text{IN}}$  should be applied before  $V_{\text{BIAS}}$  if the Slow Start feature is used.
- 2)  $C_{\text{DELAY}}$  capacitor of 10uF is adequate for  $V_{\text{BIAS}}$  rise times of up to 50ms.



**REV L: 02/18/21** 

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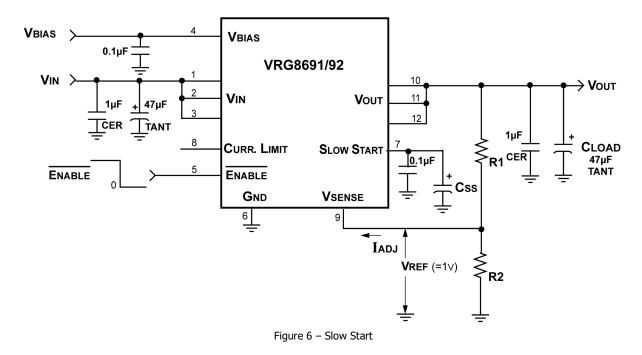
#### Application Note 4 Vout Start Up Rise Time Control

### Utilizing the Slow Start option:

When the VRG8691/92 is first powered up, using the Slow Start function controls the rate at which  $V_{OUT}$  rises to the required voltage set by R1 and R2.

#### Note:

1)  $V_{\text{IN}}$  should be applied before  $V_{\text{BIAS}}$  when the Slow Start feature is used.



If it is desirable to control the output rise time, a capacitor ( $C_{SS}$ ) can be asserted on the Slow Start pin to adjust the rise time for the following:

A. Large load capacitance will cause high surge currents which will trip the current limit circuitry.

The use of Css will allow the output voltage to rise slowly thus mitigate the surge current phenomenon.

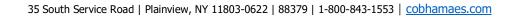
$$\frac{C_{SS}}{C_{LOAD}} > \frac{V_{OUT NOM}}{I_{CL} - I_{LOAD NOM}} X (0.4 \times 0.0014)$$

B. Css may be used solely to control VOUT RISE TIME (Tr), when CLOAD is not an issue.

$$T_R = \left(\frac{C_{SS} \times 2.5}{0.0014}\right)$$
 Note:  $C_{SS}$  in Farads and  $T_R$  in seconds.

**C.** $C_{SS}$  is effective only when  $V_{IN}$  is applied prior to  $V_{BIAS}$  or  $\overline{ENABLE}$ .

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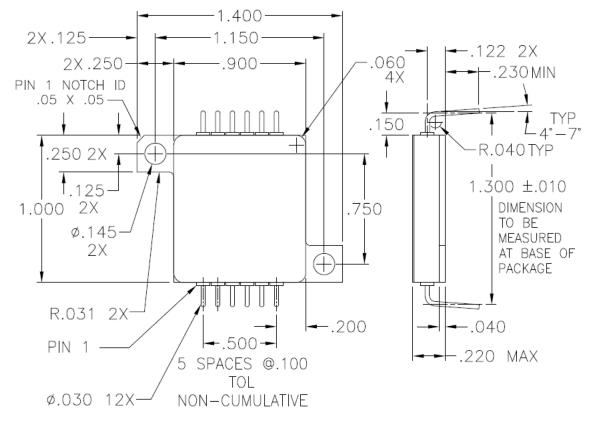


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### **Pin Numbers vs Function**

Pin	Function	Pin	Function
1	V <sub>IN</sub>	7	Slow Start
2	V <sub>IN</sub>	8	Current Limit
3	V <sub>IN</sub>	9	V <sub>SENSE</sub>
4	V <sub>BIAS</sub>	10	V <sub>OUT</sub>
5	ENABLE	11	Vout
6	GROUND	12	Vout



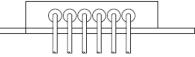


Figure 7– Package Outline — VRG8691 Thru-Hole Power Package

#### Notes:

- 1) Dimension Tolerance: ±.005 inches
- 2) Package contains BeO substrate
- 3) Case electrically isolated

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**REV L: 02/18/21** 

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### **Pin Numbers vs Function**

Pin	Function	Pin	Function
1	V <sub>IN</sub>	7	Slow Start
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4	VBIAS	10	Vout
5	ENABLE	11	Vout
6	GROUND	12	Vout

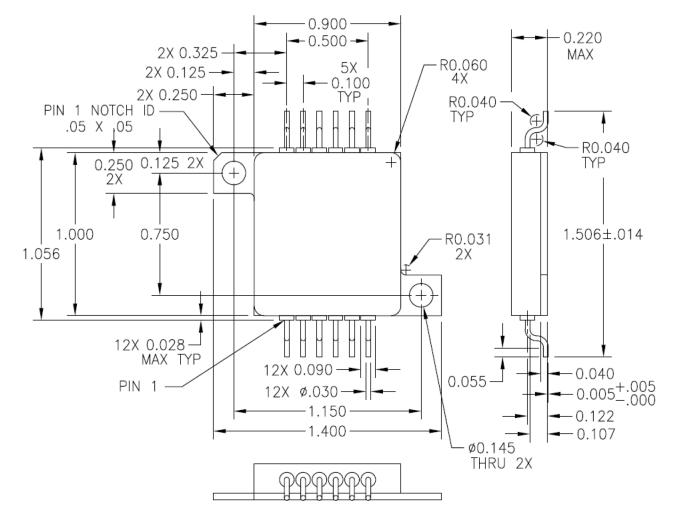


Figure 8– Package Outline – VRG8692 Surface Mount Power Package

Notes:

- 1) Dimension Tolerance: ±.005 inches
- 2) Package contains BeO substrate
- 3)Case electrically isolated





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### **Ordering Information**

Model	DLA SMD #	Screening	Package
VRG8691 - 7	-	Commercial Flow, +25°C testing only	
VRG8691 - S	-	Military Temperature, -55°C to +125°C Screened in accordance with the individual Test Methods of MIL-STD-883 for Space Applications	12 Lead
VRG8691- 201-1S	5962-0923701KXC	In accordance with DLA SMD	Thru-Hole Power Pkg
VRG8691- 201-2S	5962-0923701KXA		FOWER FKg
VRG8691- 901-1S	5962R0923701KXC	In accordance with DLA Certified RHA	
VRG8691- 901-2S	5962R0923701KXA	Program Plan to RHA Level "R", 100 krad(Si)	
VRG8692 - 7	-	Commercial Flow, +25°C testing only	
VRG8692 - S	-	Military Temperature, -55°C to +125°C Screened in accordance with the individual Test Methods of MIL-STD-883 for Space Applications	12 Lead
VRG8692- 201-1S	5962-0923701KYC	In accordance with DLA SMD	Surface Mount
VRG8692- 201-2S	5962-0923701KYA		Power Pkg
VRG8692- 901-1S	5962R0923701KYC	In accordance with DLA Certified RHA Program Plan to RHA	
VRG8692- 901-2S	5962R0923701KYA	Level "R", 100 krad(Si)	

### **Revision History**

Date	Revision	Change Description
03/24/2016	J	Import into CAES format
02/18/2021	L	REVISED PER ECN 23515





## SCD8692 7.5 Amp LDO Adjustable Positive Voltage Regulators

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### Datasheet Definitions

	DEFINITION
Advanced Datasheet	CAES reserves the right to make changes to any products and services described herein at any time without notice. The product is still in the development stage and the datasheet <b>is subject to change</b> . Specifications can be <b>TBD</b> and the part package and pinout are <b>not final</b> .
Preliminary Datasheet	CAES reserves the right to make changes to any products and services described herein at any time without notice. The product is in the characterization stage and prototypes are available.
Datasheet	Product is in production and any changes to the product and services described herein will follow a formal customer notification process for form, fit or function changes.

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